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Abstract: This paper presents a novel algorithm to emulate the Rician K-factor of a wireless communication environment using a reverberation chamber. In particular, it was focused on the frequency bands used by 5G technology for data transmission. Unlike the current state of the art, the increase of the Rician K-factor is not achieved by inserting lossy material in the chamber, but through the proposed algorithm, which selects a subset of electromagnetic field realizations inside the chamber. The algorithm was first applied to the results of a simulation obtained by using an analytical model that predicts the main physical quantities inside a rectangular cavity where a multiple monopole source stirring action is implemented. Subsequently, the robustness of the algorithm and its applicability was tested on a set of data retrieved from experimental measurements. In all of the considered scenarios, the validity of the proposed method was demonstrated.

Keywords: Rician K-factor; reverberation chamber; wireless communications

1. Introduction

The 5G network represents the fifth generation of cellular technology. It was designed to increase speed, reduce latency and improve the flexibility of wireless services. Currently, 5G technology has a theoretical peak speed of 20 Gbps, while the peak speed of 4G is only 1 Gbps [\[1\]](#page-11-0).

The antenna system incorporates the Massive MIMO (multiple input, multiple output) concept, which allows multiple transmitters and receivers to transfer more data simultaneously and in different directions, increasing the complexity of the channel model.

Furthermore, 5G New Radio [\[2\]](#page-11-1), the global standard for a more suitable 5G wireless air interface, covers spectrums not used in 4G. As an example, in Italy, the frequencies assigned to 5G are around 700 MHz, 3.5 GHz and 26 GHz. However, 5G technology is not limited to the New Radio spectrum. It is designed to support heterogeneous networks that combine both licensed and unlicensed wireless technologies. A complete overview of 5G technology can be found in [\[3–](#page-11-2)[6\]](#page-11-3).

Considering all the novelties introduced in 5G technology, the necessity arises to test these new and complex systems in a controlled environment that reproduces the propagation channels for these new frequencies.

The propagation channel of a wireless system is usually described through its statistical properties. The Rician *K*-factor is one of the main indicators used to describe the wireless channel. In fact, it represents the ratio of the signal power in the line-of sight (LOS) component over the scattered power. The *K*-factor is important for link budget calculations [\[7,](#page-11-4)[8\]](#page-11-5), or in optimizing the modulation and coding schemes of a given channel [\[9](#page-11-6)[,10\]](#page-11-7).

The 5G propagation environments [\[11\]](#page-11-8) are characterized by very high values of the Rician K-factor in urban areas (82 according to [\[12\]](#page-11-9)) and a lower value (3 to 9) for suburban, rural, and indoor environments [\[7,](#page-11-4)[12,](#page-11-9)[13\]](#page-11-10).

Finally, it should be noted that other parameters can also be adopted to characterize a propagation channel emulated inside an RC, such as the time delay spread and the

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coherence bandwidth [\[14\]](#page-11-11). These parameters are typically tuned by adding absorbing materials [\[15\]](#page-11-12), with a consequent effect on the K-factor. The possibility of tuning the K-factor regardless of other parameter settings represents an advantage in using the RC for testing wireless systems.

In order to perform a test on 5G equipment before their installation in a real scenario, preliminary tests can be performed in a more controlled test site, such as within Reverberation Chambers (RCs), which are able to provide multipath propagation conditions.

RCs are usually characterized by lower values of Rician K-factor [\[16,](#page-11-13)[17\]](#page-12-0); in fact, a well stirred RC has a high value of stirred components of the electromagnetic field and low values of its unstirred ones. In order to increase the Rician K-factor to emulate wireless propagation environments, lossy elements are usually inserted. As a consequence, the quality factor of the RC decreases; this means that the strength of the electromagnetic fields become lower and, consequently, a higher amplification in the entrance to the chamber is needed. This implies a higher cost of hardware equipment.

The aim of this paper is to enhance the value of the Rician K-factor by properly selecting the field realizations inside the RC. In this way, the quality factor of the RC will not be reduced, since no further lossy elements are inserted.

The traditional methods available to increase the K-factor exhibit some complexity. In fact, the reduction of the RC quality factor surely acts on the K-factor [\[18\]](#page-12-1) because the scattered waves are attenuated when they intercept the lossy objects located in the RC. As a consequence, the final effect might depend on the number and positioning of the absorbers with respect to the antennas and stirrers [\[19\]](#page-12-2). Furthermore, the direct component should not be affected by the absorber's positioning, therefore, a pure line-of-sight positioning between the transmitting and the receiving antenna is required. Last, but not least, the amount of direct component also depends on the directivity of the antennas [\[20\]](#page-12-3); so, making the antenna orientation another important variable acting on the K-factor.

The proposed method is applicable to RCs where the field configurations are known a priori and repeatable, such as mechanical stirring using one or more metallic paddles in step mode, or the Oscillating Walls Stirred (OWS) RC [\[21\]](#page-12-4), or, as in our case, to the Multiple Monopole Source Stirring RC. This particular implementation of a source stirring technique is achieved thanks to an array of monopole antennas mounted on the walls of the RC and, subsequently, fed to obtain the stirring action [\[22\]](#page-12-5). This technique was successfully compared to the more classical approach implementing both mechanical stirring and MMSS in the same chamber [\[23\]](#page-12-6). This paper presents an analytical and an experimental approach for demonstrating the validity of the proposed method.

The paper is organized as follows: Section [2](#page-1-0) describes the scenario inside an RC to simulate an urban environment; Section [3](#page-3-0) describes the rationale of our proposal and the corresponding algorithm; Section [4](#page-6-0) shows the results; Section [5](#page-10-0) deals with the discussion of the results and concludes the paper.

2. Scenario

The considered scenario is an RC where Multiple Monopole Source Stirring (MMSS) is implemented. The RC has dimensions of 800 mm \times 900 mm \times 1000 mm and is made of galvanized steel. The stirring technique is obtained using a certain number of monopoles placed on the chamber's wall, however in our case, for practical reasons, the stirring of the chamber is performed by inserting a single monopole each time in one of the 120 holes present on the walls of the RC, 20 holes in each face. In this way, the reverberation effect is due to 120 different configurations of the electromagnetic field inside the cavity.

Owing to the reverberating characteristic of the RC, the field inside the chamber has statistical properties that, under proper preparation of the chamber, resembles the typical propagation channel of an urban environment. In particular, the environmental channel has statistical properties that follows a Rician distribution with a specific K-factor depending on several factors (antenna gain, elevation, etc.) [\[7–](#page-11-4)[10\]](#page-11-7).

Placing a dipole antenna inside the MMSS chamber, the insertion loss *S*²¹ among each monopole on the walls and the dipole can be evaluated; hence, it is possible to predict the performance of the RC in terms of its statistical indicators, in particular, we focused on the Rician K-factor (K) and the quality factor (Q,).

The Rician K-factor of the chamber [16] [can](#page-11-13) be computed according to (1):

$$
K = \frac{(|\langle S_{21} \rangle|)^2}{\langle |\mathcal{S}_{21} - \langle \mathcal{S}_{21} \rangle|^2 \rangle},\tag{1}
$$

where $\langle \cdot \rangle$ means averaging the operator over all the chamber field realizations.

The quality factor is computed according to (2) [\[24\]](#page-12-7): The quality factor is computed according to \mathcal{I}_1

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$$
Q = \frac{16\pi^2 V}{\eta_{TX}\eta_{RX}\lambda^3} \left\langle \frac{|S_{21}|^2}{1 - |S_{11}|^2} \right\rangle
$$
 (2)

where V is the chamber volume, λ is the wavelength (m), η_{TX} and η_{RX} are the total antenna where \vec{v} is the enancer volume, \vec{v} is the wavelength $\langle \text{m} \rangle$, $\eta_1 \chi$ and η_{RX} are efficiency factors for the transmit and receive antennas, respectively. The same scenario instantant with the MMSS chamber can be determined with an analytical application and the model

The same scenario inside the MMSS chamber can be modeled with an analytical approach $[14]$, so that Equations (1) and (2) can be applied to a simulated S_{21} parameter. The limitation of this approach is due to the presence of losses from different sources (walls, antennas, apertures, gaskets) that cannot be modelled as is. For this reason, a classical approach in RC chamber [\[22\]](#page-12-5) modelling is used: the chamber's walls are simulated with equivalent conductivity that permits us to obtain a chamber with the same Q-factor as the real one, for each frequency $[25]$. In particular, in our chamber, in order to obtain the same Q-factor, the conductivity of the walls is modelled with a value that is reduced by a factor of 10 with respect to the σ of the galvanized steel. The limitation of this approach is due to the presence of losses from different sources (α)

If the RC is well stirred, looking at the scatter plot of S_{21} obtained with simulations of the MMSS chamber, the points that represent the complex value of the field in a point inside the chamber, are almost distributed around the origin of the axis and almost uniformly distributed on the four quadrants (Figure [1,](#page-3-1) left). As a consequence, the K-factor is low because the numerator of Equation (1) is close to zero. In order to increase the K-factor, and consequently, obtain an environment similar to the urban environment, the traditional approach introduces losses into the RC. In this way, the Q factor decreases and the scatter plot of *S*₂₁ p[re](#page-3-1)sents a more evident bias and, consequently, K also increases (Figure 1, right), and the strength of the electromagnetic field in the chamber decreases.

Figure 1. *Cont*.

Figure 1. Scatter plot of S_{21} and an indication of the Q and K parameters related to an empty RC (**up**) and the same RC with the insertion of lossy elements (**down**). (**up**) and the same RC with the insertion of lossy elements (**down**).

3. The Proposed Method 3. The Proposed Method

In order to increase the K-factor of the chamber, without reducing the Q-factor, a new approach is proposed. The idea is to select the proper field configurations inside the ber that have an ensemble statistical property that follows the multipath fading urban chamber that have an ensemble statistical property that follows the multipath fading urban environment property. For this reason, the method is applicable to a stirring technique that that permits us to obtain repeatable and well controlled field configurations. permits us to obtain repeatable and well controlled field configurations.

Figure 2 [sh](#page-4-0)ows the idea. In Figure 2, [th](#page-4-0)e scatter plot of an ideal reverberation chamber Figure 2 shows the idea. In Figure 2, the scatter plot of an ideal reverberation chamber is shown. In particular, the blue points represent the S_{21} for 5000 field configurations inside an ideal chamber. The number of configurations is high in order to enhance the statistical properties and obtain a better visualization of the idea. It can be seen that the blue points are uniformly distributed around the center of the scatter plot and describe almost a perfect circle around it. In this case, the K-factor is almost zero, and the maximum radius of the circle is 1. This configuration does not have any bias and so cannot represent an urban environment well. However, by selecting the proper configurations, it is possible to obtain the desired statistical properties. This is explained in the selected coloring configuration of Figure [2](#page-4-0) shown in yellow, purple or orange. The subintervals of the figures are represented by three different possible selections of field configurations. The yellow one is obtained by choosing samples that are within a circle of radius R = 0.3 centered in point C(−0.4, 0.4). The center of the circle represents the average values of the chosen samples (the numerator
The center (1)) subseque Parameter the average values of the consultation of the consultation of the of Equation (1)), whereas R represents the maximum distance in the complex plane of the complex plane of the similar state of the maximum distance in the complex plane of the maximum distance of the similar state of the s samples miside the circle and the average values of the circle (the distance of each sample represents the denominator of Equation (1) before averaging). For the other configurations, distance of each sample represents the denominator of Equation (1) before averaging). For the orange one is chosen by selecting samples that have the same average values of the yellow one, but that are closer to the average value $(R = 0.2, C(0.4, 0.4))$. In this case, the K-factor is 16.13, greater than the yellow one that has $K = 6.77$. The purple one is chosen by 1) in this case, the K-factor is 16.13, greater than the yellow one that the same that have the same distance with their average value as the orange one, but have a lower average value ($R = 0.2$, C(0.2, -0.2)). In this case, K = 3.8. samples inside the circle and the average values of the circle (the distance of each sample

This simple example demonstrates that in order to improve the K-factor, it is necessary both to increase the average value of the chosen samples and to reduce the distance between the samples and their average. However, in the example, a very large number of field configurations are available for the selection, so more than one selection ca be obtained with the same K-factor. On the other hand, in a real chamber environment, the number of samples are lower, and not uniformly distributed so well. For this reason, the implementation of the idea optimizes the selection in order to obtain a subinterval with the maximum number of field configurations.

Figure 2. $\frac{1}{2}$ in an ideal reverber plot of the indicate $\frac{1}{2}$ in an ideal represents and purple points represent the coloridation of the field. The vollow areas and purple points represent the coloridation the possible configurations of the field. The yellow, orange and purple points represent the selected the selected of the sele configurations for different values of R and C, giving a different value of K-factor. configurations for different values of R and C, giving a different value of K-factor. **Figure 2.** Scatter plot of the *S*²¹ in an ideal reverberation chamber. The blue point represents all of

In are following, are imprementation of the fact applied to a real chamber is accelerated.
In particular, the MMSS chamber is used to demonstrate the feasibility of the method. In the following, the implementation of the idea applied to a real chamber is described.

In order to make the selections, it is necessary to obtain all of the possible configurations inside the specific chamber. In particular, the S_{21} paramater is measured using, for example, a Vector Network Analyzer, or is calculated analytically using the theoretical model [22] Inside the MMSS chamber the S₂₅ is evaluated between each monopole on the model [\[22\]](#page-12-5). Inside the MMSS chamber, the *S*₂₁ is evaluated between each monopole on the walls of the chamber walls of the chamber.

The selection algorithm proposed is the following:

- 1. The desired Rician K-factor, K^* , is chosen.
- In the following, the implementation of the idea applied to a real chamber is dependent of the independent of \mathbf{r} 2. For each i-th S_{21} sample, the normalized distance with all the others in the complex plane is computed:

$$
dist(i,k) = \frac{\left| S_{21}^{(i)} - \left| S_{21}^{(k)} \right| \right|}{\left| S_{21}^{(i)} \right|},\tag{3}
$$

- 3. A threshold for *dist* called *dist*_{*MAX*} is set to 1. This threshold corresponds in Figure [2](#page-4-0) to the radius R of the selection normalized to the amplitude of the i-th *S*₂₁.
- 4. For each i-th *S*₂₁ sample, all of the other *S*₂₁ that satisfies the condition $dist(i, k) < dist_{MAX}$ are selected This subinterval of samples represents the neighbors of i-th *S*₂₁ having normalized the distance to lower than *dist_{MAX}* and can be indicated as *S*21_*near*(*i*, *dist*_{*MAX*}) and their number of elements indicated as $N_{dis}^{(i)}$
Among all of the *S*21 *near*(*i distance*) subintervals, the one (*i*^{*}) that h *distMAX* .
- 5. Among all of the $S21$ *near*(*i*, *dist_{MAX}*) subintervals, the one (i^{*}) that has the largest nd the Rician K-factor related to this subset is number of element $N_{dis}^{(i)}$ $\frac{d}{dist_{MAX}}$ is chosen, and the Rician K-factor related to this subset is computed, K(i*).

6. If $K(i^*)$ is lower than K^* the algorithms reiterate from point 5, increasing $dist_{MAX}$ by 0.1 until the condition $K(i^*)$ > K^* is reached.

 $\overline{}$

7. Once the desired K-factor is obtained, the procedure stops, and the last subinterval 7. Once the desired K-factor is obtained, the procedure stops, and the last subinterval identified is the selection sought. identified is the selection sought.

For a more immediate intelligibility, a flowchart describing the algorithm is given in For a more immediate intelligibility, a flowchart describing the algorithm is given in Figure 3. Figur[e 3](#page-5-0).

Figure 3. Flowchart describing the proposed algorithm. **Figure 3.** Flowchart describing the proposed algorithm.

The proposed algorithm requires two considerations. The first one is that the greater the number of field configurations, the more efficient the proposed solution. This means that the more are the elements in the field realization subset that have the desired Rician K-factor, then the better are the statistics in terms of electromagnetic stirring.

The second consideration is that not all of the values of K^* can be reached; however, the next section shows that by considering the values of K^* , which represent wireless propagation environments according to the literature [\[8](#page-11-5)[,9\]](#page-11-6), the algorithm returns successful results.

4. Results Ω esults section, the application of the proposed algorithm is reported. In particular, Ω

In this section, the application of the proposed algorithm is reported. In partic-ular, Section [4.1](#page-6-1) reports on the analytical simulations and Section [4.2](#page-7-0) reports on the experimental data. *4.1. Analytical Results*

4.1. Analytical Results

Considering the MMSS chamber described in Section 2, the S_{21} parameters between the dipole and the 120 monopoles on the walls are evaluated with the analytical model explained in [\[22\]](#page-12-5). The model is based on a modal approach and all of the possible 120 field configurations, corresponding to each monopole used as feeders of the chamber, are
undertial. Figure 4 shows the scatter plot of the 21 computed at the frequency of 750 MHz; evaluated. Figure [4](#page-6-2) shows the scatter plot of the *S*₂₁ computed at the frequency of 750 MHz;
the corresponding value of the Rician K-factor is 0.18. As expected, the chamber is well the corresponding value of the Rician K-factor is 0.18. As expected, the chamber is well stirred so the K-factor is low. stirred so the K-factor is low.

N_f = 120, K = 0.18, f = 750 MHz

Figure 4. Scatter plot of S_{21} relative at a 120 electromagnetic field realization, simulated at 750 MHz. The corresponding value of the Rician K-factor is 0.18. The corresponding value of the Rician K-factor is 0.18.

If we want to increase the Rician K-factor to a value of 7, to simulate a rural environ-ment of wireless propagation, we can apply the proposed algorithm. If we want to increase the Rician K-factor to a value of 7, to simulate a rural environ-

Starting from $dist_{MAX} = 1$, the iterative procedure lasts with a value of $dist_{MAX} = 2.7$. The procedure stops after obtaining the electromagnetic field realization subset that has a Rician K-factor value of 7.24, as shown in Figure [5.](#page-7-1)

The figure also shows the number of field configurations necessary to obtain the desired statistics: in this example, 19. It is important to highlight that the procedure is not limited to obtaining the number of field configurations, but it sets, exactly, which are the configurations to be used. In the MMSS chamber, this corresponds to setting which are the specific monopoles to switch on during the tests.

 $N_f = 19$, K = 7.24, f = 750 MHz

Figure 5. Scatter plot of S_{21} relative at the subset of 19 electromagnetic field realizations simulated at 720 MHz. The assument diagonalized at V factor is 7.24 at 750 MHz. The corresponding value of the Rician K-factor is 7.24. 750 MHz. The corresponding value of the Rician K-factor is 7.24.

4.2. Experimental Results

The experimental validation of the proposed method is applied to the same chamber used for the simulations. In this case, two bands reserved to $\frac{1}{5}G$ data transmission are investigated. In particular, the band between 713 and 778 MHz and the one between 3.27 and 3.80 GHz are used in Italy for the data Frequency Division Duplex (FDD) uplink and downlink [26].

insert the feeding monopoles each time. The figure also shows the transmitting 100 mm monopole and the receiving discone antenna: both antennas were designed and built in used for the simulations. In this case, two bands reserved to 5G data transmission are The MMSS RC used is shown in Figure [6.](#page-8-0) It is possible to note the holes used to our laboratories.

I igute 7 shows the placement of the transmitting and the receiving antennas. The transmitting monopole is manually placed in all of the 120 holes present in the walls of the rand 3.80 GHz are used in Italy for the data Frequency Division Duplex (Figure 2.6 GHz) chamber. The receiving antenna is placed in the working volume of the RC, a subvolume downlink [26]. spaced a quarter of a wavelength from the walls. Figure [7](#page-8-1) shows the placement of the transmitting and the receiving antennas: the

A VNA was used to measure the S_{21} parameter in 1601 frequency points, equally spaced in both frequency ranges. The goal was to select, for each frequency, the subset of chamber field realizations that have a corresponding Rician K-factor greater than 3. Figure 8 shows the Rician K-factor frequency behavior, considering the whole number of possible configurations (120), and after having applied the algorithm.

Figure 9 shows the nu[m](#page-9-1)ber of field realizations selected by the algorithm corresponding to the K-factor, shown in red in Figure 8, for each frequency i[n t](#page-9-0)he considered frequency range.

Figure 6. Experimental Setup: on the left, the reverberation chamber, and on the right, the discone and monopole antennas. and monopole antennas. and monopole antennas.

Figure 7. Insertion of the transmitting monopole (left) and the location of the receiving antenna (right) inside the RC. Each hole, clearly visible from the inner view, is subsequently used to insert the transmitting monopole toward the chamber inside.

configurations (120), and after having applied the algorithm.

Figure 8. Rician K-factor frequency behavior with the 120 field configurations and after having applied the algorithm in the frequency band 711–778 MHz.

Figure 9. Number of field realizations selected by the algorithm in the frequency band 711–778 MHz.

range was to have K equal to 5. The same procedure was applied to the higher frequency range, and the corresponding results are shown in Figures [10](#page-10-1) and [11.](#page-10-2) For the sake of demonstrating the applicability of the proposed method to different values of Rician K-factor, the goal for this frequency

range was to have K equal to 5.

Figure 10. Rician K-factor frequency behavior with the 120 field configurations and after having applied the algorithm in the frequency band 3.27-3.80 GHz. applied the algorithm in the frequency band 3.27–3.80 GHz.

Figure 11. Number of field realizations selected by the algorithm in the frequency band 3.27–3.80 **Figure 11.** Number of field realizations selected by the algorithm in the frequency band 3.27–3.80 GHz.

Figure 11. Figure 11. *Number of figures in the algorithm* **in the selection of** α **. ***Number of algorithm* in the frequency band 3.27–3.800 in the frequency band 3.27–3.800 in the frequency band 3.27–3.800 in the f GHz*.* **5. Discussion and Conclusions**

The proposed method, used to tune the Rician K-factor by selecting a subset of electromagnetic field realizations, is useful for emulating the statistical properties of the propagation environment for 5G wireless communications. The method was applied to a reverberation chamber where the multiple monopole source stirring technique was implemented, but in the theory, it can be used for any chamber where the field configurations are uniquely determinable, such as mechanical stirring in tuned mode or OWS RCs.

Its reliability and robustness were demonstrated by using both the prediction of an analytical model and the data retrieved from experimental measurements, using different values of the Rician K-factor as a target. Future work will focus on the use and programming of an electronic switch to manage the feeding monopoles in order to automatize the procedure, which in the current state of the art is manually managed, thus reducing the testing time.

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